

# Wireless Sensor Networks

## 2. Multiplexing

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# Direct Sequence Spread Spectrum

- A chip is a bit sequence (given by  $\{-1, +1\}$ ), which encode a smaller set of symbols
- E.g. Transmission signal:  $0 = (+1, +1, -1)$ ,  $1 = (-1, -1, +1)$

|          |          |          |          |
|----------|----------|----------|----------|
| 0        | 1        | 0        | 1        |
| +1 +1 -1 | -1 -1 +1 | +1 +1 -1 | -1 -1 +1 |

- Coding by calculating the inner product  $c_i s_i$  of the received signal and the chip  $c_0 = -c_1$ :

$$\sum_{i=1}^m c_{0,i} s_i \qquad \sum_{i=1}^m c_{1,i} s_i$$

- In the case of a superimposed signal, the original signal can be decoded by filter
- DSSS is used by GPS, WLAN, UMTS, ZigBee, Wireless USB based on the **Barker code**
  - Here for all  $v < m$

$$\left| \sum_{i=1}^{N-v} a_i a_{i+v} \right| \leq 1$$

- Barker Code for 11Bit: +1 +1 +1 -1 -1 -1 +1 -1 -1 +1 -1

- CDMA (Code Division Multiple Access)
  - e.g. GSM (Global System for Mobile Communication)
  - or UMTS (Universal Mobile Telecommunications System)
- Uses chip-sequence with
  - $C_i \in \{-1, +1\}^m$
  - $-C_i$  =  $(-C_{i,1}, -C_{i,2}, \dots, -C_{i,m})$
- so that the normalized inner product for all  $i \neq j$  the result is 0.

$$C_i \bullet C_j = \frac{1}{m} C_i \cdot (C_j)^T = \frac{1}{m} \sum_{k=1}^m C_{i,k} C_{j,k} = 0.$$

$$\begin{array}{cccc} +1 & +1 & -1 & -1 \\ -1 & +1 & -1 & +1 \end{array}$$

$$-1 + 1 + 1 - 1 = 0$$

- Synchronized recipients get a linear combination of A and B
- Multiplying by the desired chip sequence yields the desired message.

$$B \bullet C = 0$$

$$C \bullet C = 1$$

$$C \bullet (-C) = -1$$

$$(A + \cancel{B}) \bullet C = A \bullet C + \underbrace{B \bullet C}_0$$

$$(A + 2 \bullet B) \bullet C = \underline{A \bullet C}$$

# CDMA: Example 1

$$(2 \cdot C_1 + 3 \cdot C_2) \cdot C_1 = 2 \cdot \overset{-1}{\cancel{C_1}} + 3 \cdot \overset{-1}{\cancel{C_2}} = 2 \cdot (-1) + 3 \cdot (-1) = -5$$

$$(-5, 1) \cdot (+1, -1) = \frac{6}{2} = 3$$

$$C_1 \otimes C_2 = \begin{pmatrix} (-1) \\ 0 \\ (-1) \end{pmatrix} + \begin{pmatrix} (-1) \\ 0 \\ (+1) \end{pmatrix} = \begin{pmatrix} -2 \\ 0 \\ 0 \end{pmatrix}$$

+1     +(-1)

$$C_1 \otimes (-C_2) = \begin{pmatrix} (-1) \\ 0 \\ (-1) \end{pmatrix} + \begin{pmatrix} (+1) \\ 0 \\ (-1) \end{pmatrix} = \begin{pmatrix} 0 \\ 0 \\ -2 \end{pmatrix}$$

$$(2) \cdot C_1 + (3) \cdot C_2 = (-2-3, -2+3) = (-5, 1)$$

$$(-5, 1) \cdot \begin{pmatrix} (-1) \\ (-1) \end{pmatrix} = \frac{5 \cdot 1}{2} = 2.5 \approx 2$$

- Sender A:

- 0 = (-1, -1) =  $C_1$

- 1 = (+1, +1)

- Sender B:

- 0 = (-1, +1) =  $C_2$

- 1 = (+1, -1)

- A sends 0, B sends 0:

- Result: (-2, 0)

- C receives (-2, 0):

- Decoding of A:  $(-2, 0) \cdot (-1, -1) = (-2)(-1) + 0(-1) = 2$

- A has therefor sent 0 because result is positive

# CDMA: Example 2

- Sample-code:
  - Code  $C_A = (+1, +1, +1, +1)$
  - Code  $C_B = (+1, +1, -1, -1)$
  - Code  $C_C = (+1, -1, +1, -1)$
- A sends Bit 0, B sends Bit 1, C sends nothing
  - $V = C_1 + (-C_2) = (0, 0, 2, 2)$
- Decoding for A:  $V \cdot C_1 = (0, 0, 2, 2) \cdot (+1, +1, +1, +1) = 4/4 = 1$ 
  - results in Bit 0
- Decoding for B:  $V \cdot C_2 = (0, 0, 2, 2) \cdot (+1, +1, -1, -1) = -4/4 = -1$ 
  - results in Bit 1
- Decoding for C:  $V \cdot C_3 = (0, 0, 2, 2) \cdot (+1, -1, +1, -1) = 0$ 
  - results in: no Signal.

$C_D ?$

- Multiplexed
  - Spatial Multiplexing
  - Frequency division multiplexing
  - Time division multiplexing
  - Code division multiplexing
  - Multiple-input multiple-output (next lecture)
- Modulation
  - Amplitude modulation
  - Phase modulation
  - Frequency modulation

- $i$ : imaginary number with
  - $i^2 = -1$
- A complex number is a linear combination of a real part  $a$  and imaginary  $b$ 
  - $z = a + bi$
- Calculation rules:
  - $(a+bi)+(c+di) = (a+c) + (b+d) i$
  - $(a+bi) (c+di) = (ac - bd) + (ad + bc) i$
  - $1/ (a+bi) = (a-bi)/(a^2+b^2)$
- Complex conjugate
  - $(a+bi)^* = (a - bi)$

- Important equation
  - $e^{i\pi} = -1$
  - $e^{i\varphi} = \cos \varphi + i \sin \varphi$
- Exponentiation of a complex number
  - $e^{a+bi} = e^a e^{bi} = e^a (\cos b + i \sin b)$
- Therefore
  - real part  $e^{i\varphi}$ :  $\operatorname{Re}(e^{i\varphi}) = \cos \varphi$
  - imaginary of  $e^{i\varphi}$ :  $\operatorname{Im}(e^{i\varphi}) = \sin \varphi$



- Real number representation

- Sine and cosine functions of different frequencies

$$g(x) = \sum_{k=0}^{N-1} a_k \cos \frac{2\pi kt}{T} + b_k \sin \frac{2\pi kt}{T}$$

- Computation of the inverse by cosine/sine integral product

$$a_k = \frac{2}{T} \int_0^T g(t) \cos(2\pi nft) dt$$

$$b_k = \frac{2}{T} \int_0^T g(t) \sin(2\pi nft) dt$$

- Complex representation

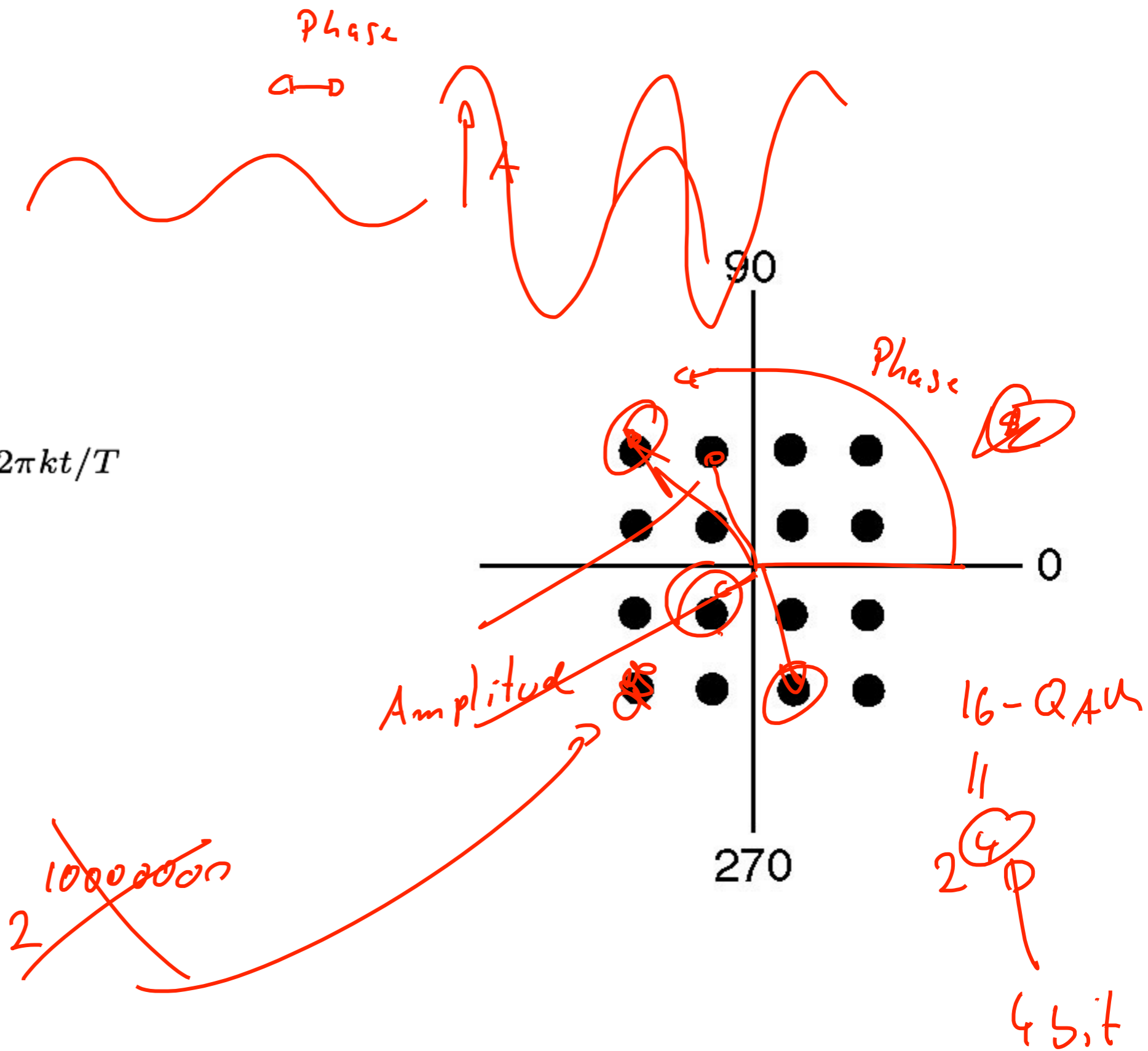
- real part of the exponential function of different frequencies

$$f(x) = \sum_{k=0}^{N-1} z_k e^{i2\pi kt/T}$$

- Computation of the inverse by the integral over the product with the complex conjugated carrier wave

$$z_k = \frac{1}{T} \int_0^T \left( e^{i2\pi kt/T} \right)^* f(x) dt$$

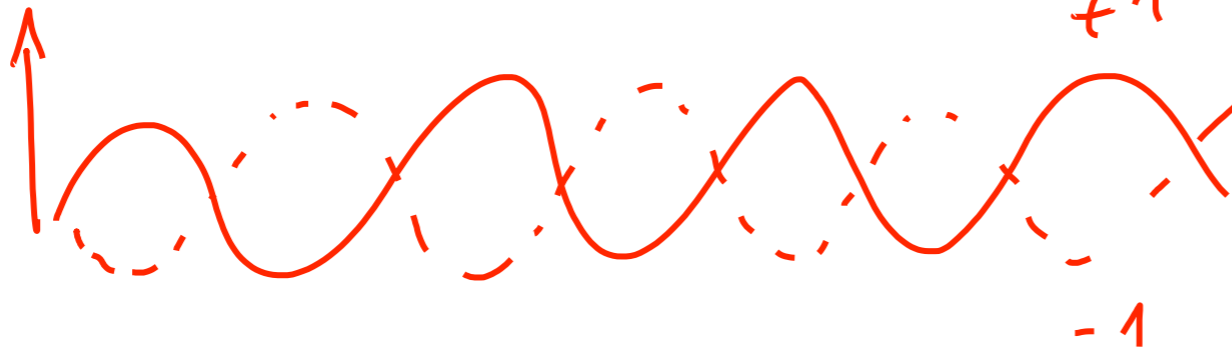
$$f(x) = \sum_{k=0}^{N-1} z_k e^{i2\pi kt/T}$$



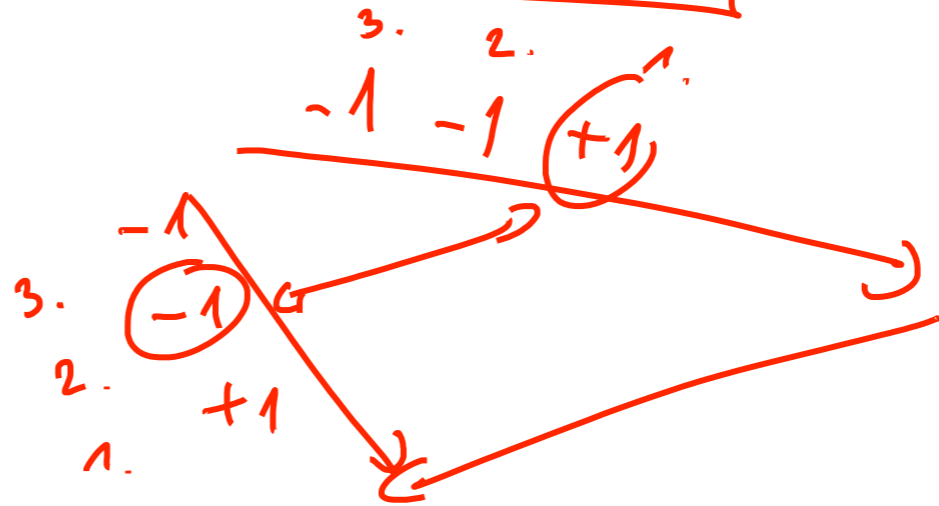
$A = +1 \quad -1$

$$\begin{array}{cccc} +1 & +1 & -1 & 0 \\ -1 & -1 & +1 & 1 \\ \hline +1 & +1 & -1 & \\ \hline +1 & +1 & -1 & = 3 \rightarrow 0 \end{array}$$

Supposition



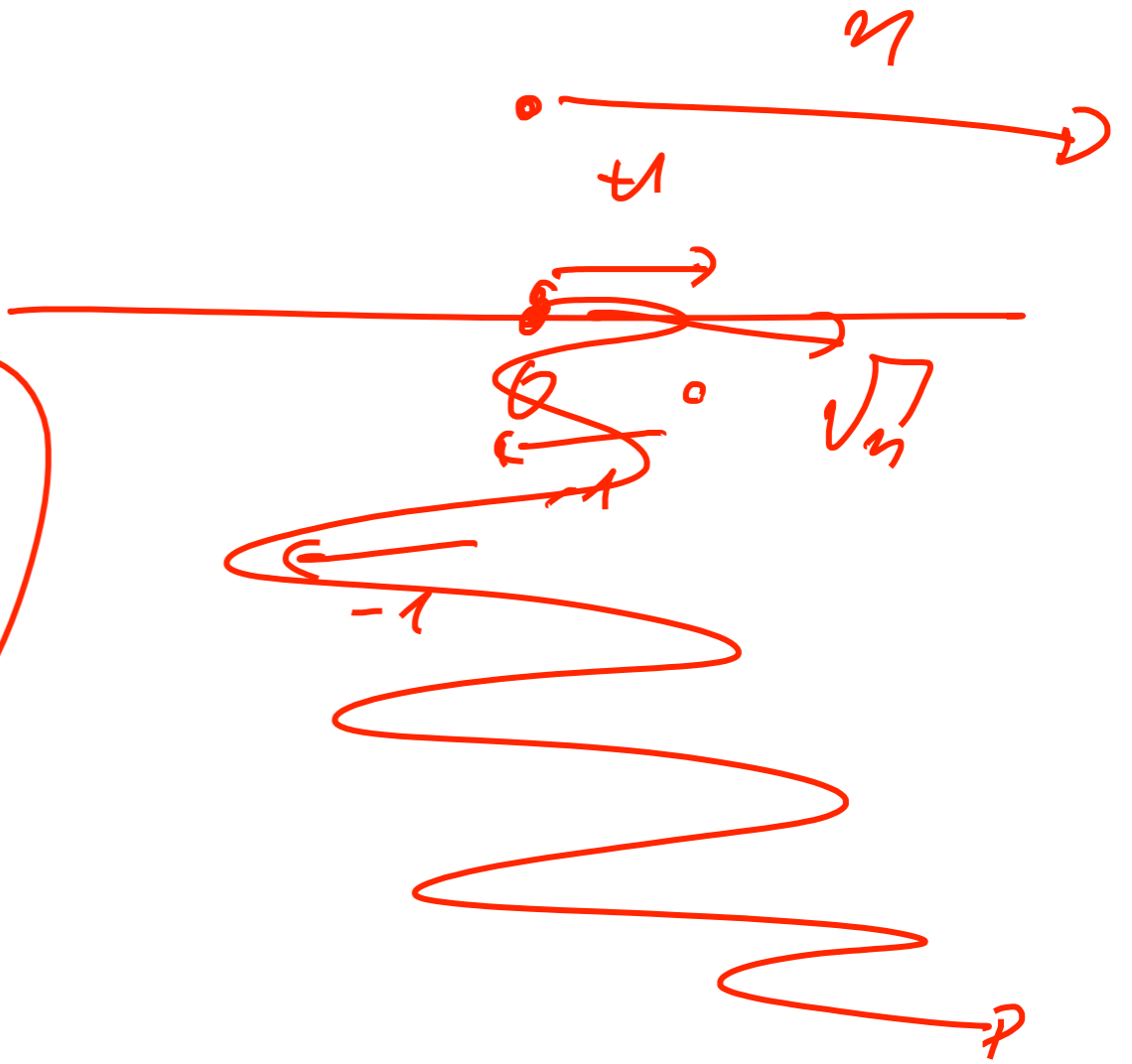
1 bit



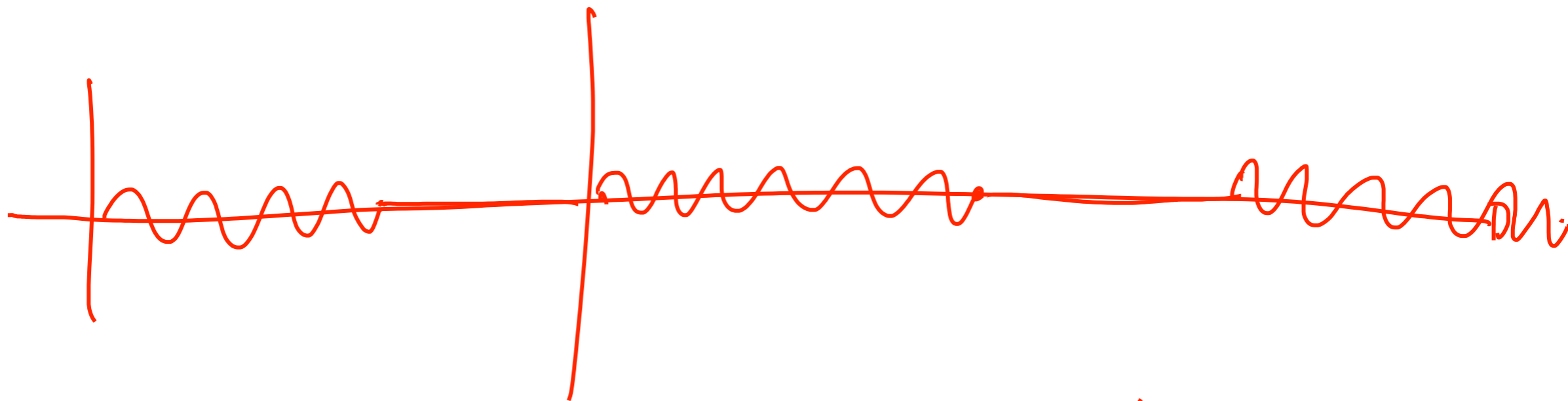


$$C \circ C = 1$$

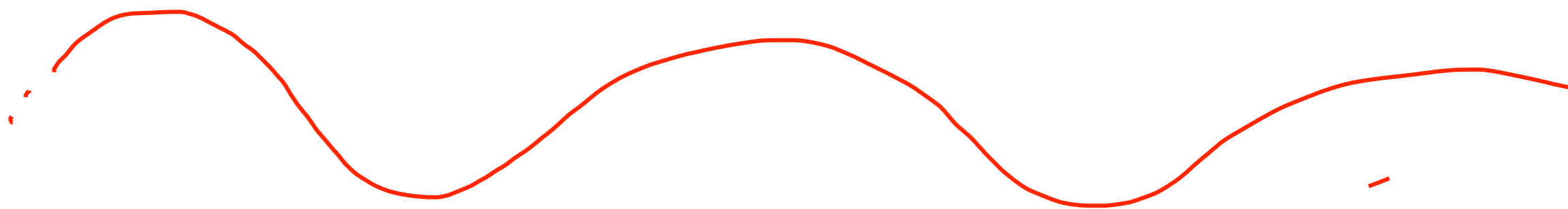
$$E[C \cdot R] = \frac{\sqrt{n}}{n} = \frac{1}{\sqrt{n}}$$



1 1 1 1 0 0 0 0 1 1 1 1 0 0 0 1 1 1 1 0 0



0 1 0 1 1 0 1 0 1 0 0 1 1 1



# Wireless Sensor Networks

## 3. Overview

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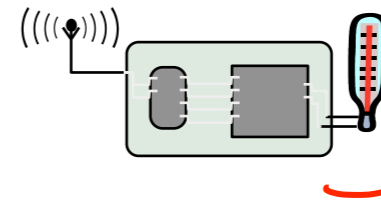
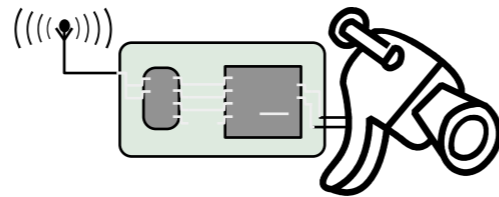
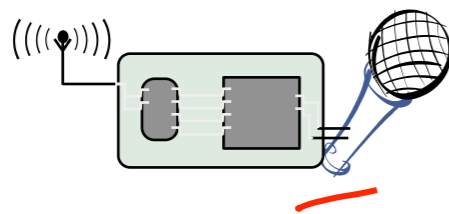
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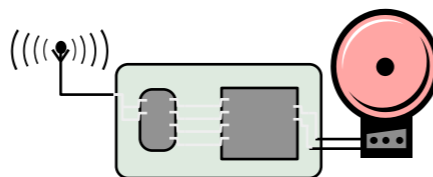
# Roles of Participants in WSN

- Sources of data: Measure data, report them “somewhere”
  - Typically equip with different kinds of actual sensors

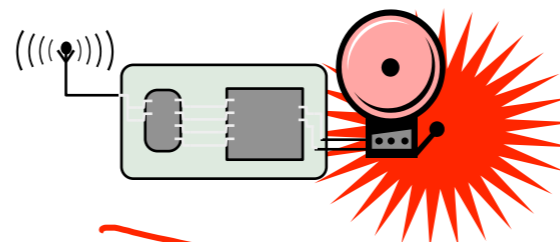
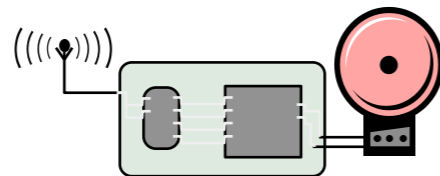


*→ gateways*

- Sinks of data: Interested in receiving data from WSN
  - May be part of the WSN or external entity, PDA, gateway, ...



- Actuators: Control some device based on data, usually also a sink





## ■ Main components of a WSN node

- Controller
- Communication device(s)
- Sensors/actuators
- Memory
- Power supply

